PERFORMANCE EVALUATION OF PAPR REDUCTION IN FBMC SYSTEM USING NON-LINEAR COMPANDING TRANSFORM

A Project report submitted in partial fulfilment of the requirements for the award of the degree of

BACHELOR OF TECHNOLOGY IN

ELECTRONICS AND COMMUNICATION ENGINEERING

Submitted by

S. Gana Venkatesh (317126512109)

Y. Renuka (318126512L18)

M. Himaja (317126512089)

G. Lavanya (3181265120L22)

Under the guidance of

Dr. A. Lakshmi Narayana

Asst.Professor-Department of ECE



DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEERING

ANIL NEERUKONDA INSTITUTE OF TECHNOLOGY AND SCIENCES (UGC AUTONOMOUS) (*Permanently Affiliated to AU, Approved by AICTE and Accredited by NBA & NAAC with 'A' Grade*) Sangivalasa, Bheemili mandal, Visakhapatnam. (A.P) 2020-2021

DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEERING

ANIL NEERUKONDA INSTITUTE OF TECHNOLOGY AND SCIENCES

(Permanently Affiliated to AU, Approved by AICTE and Accredited by NBA & NAAC with 'A' Grade)

Sangivalasa, Bheemili mandal, Visakhapatnam dist. (A.P)



CERTIFICATE

This is to certify that the project report entitled "Performance Evaluation Of PAPR Reduction in FBMC System Using Nonlinear Companding Transform" submitted by S. Gana Venkatesh (317126512109), Y. Renuka (318126512L18), M. Himaja (317126512089), G. Lavanya (318126512L22) in partial fulfilment of the requirements for the award of the degree of Bachelor of Technology in Electronics & Communication Engineering of Andhra University, Visakhapatnam is a record of Bonafede work carried out under my guidance and supervision.

Atr Project Guide

Dr. A. Lakshmi Narayana Assistant Professor, M. Tech, PHD Department of E.C.E

ANIASsistant Professor Department of E.C.E. Anil Neerukonda Institute of Technology & Sciences Sangivalasa, Visakhapatnam-531 162

Head of the Departmen

Dr.V. Rajya Lakshmi Professor & HOD Department of E.C.E

ANITS Head of the Department Department of E C E Anil Neerukonda Institute of Technology & Science Sangivalasa - 531 162

2



ACKNOWLEDGEMENT

We would like to express our deep gratitude to our project guide **Dr. A. Lakshmi Narayana** M. Tech, PhD, Department of Electronics and Communication Engineering, ANITS, for his guidance with unsurpassed knowledge and immense encouragement. We are grateful to **Dr.V. Rajya Lakshmi**, Head of the Department, Electronics and Communication Engineering, for providing us with the required facilities for the completion of the project work.

We are very much thankful to the **Principal and Management, ANITS, Sangivalasa**, for their encouragement and cooperation to carry out this work.

We express our thanks to all teaching faculty of the Department of ECE, whose suggestions during reviews helped us in accomplishment of our project. We would like to thank all non-teaching staff of the Department of ECE, ANITS for providing great assistance in accomplishment of our project.

We would like to thank our parents, friends, and classmates for their encouragement throughout our project period. At last, but not the least, we thank everyone for supporting us directly or indirectly in completing this project successfully.

PROJECT STUDENTS

S. Gana Venkatesh (317126512109), Y. Renuka (318126512L18), M. Himaja (317126512089), G. Lavanya (318126512L22).

ABSTRACT

Filter Bank Multicarrier with Offset Quadrature Amplitude Modulation (FBMC/OQAM) is one of the best waveform used in 5G communications. The FBMC is a multicarrier modulation system (MCM) which suffers from high peak to average power ratio (PAPR) causing a reduction in power efficiency. Nonlinear companding techniques are widely used to reduce the peak-to-average-power ratio (PAPR). In this paper, we have evaluated the performance of different non-linear companding techniques to reduce the PAPR in FBMC/OQAM system while achieving good Bit Error rate. We assessed the A-law, and Mu-law companding techniques. It is found that the best companding function is the Mu-law for Mu-law companding, results in greater reduction of PAPR by about 9.2336dB when compared with the original FBMC system at 10^{^-3} complementary cumulative distribution function (CCDF) and for A-law companding, the PAPR is reduced by about 3.8731 dB at 10^{^-3} of the complementary cumulative distribution function (CCDF).

CONTENTS

LIST OF FIGURES	viii
LIST OF TABLES	viii
LIST OF ABBREVATIONS	ix
CHAPTER 1: INTRODUCTION	
1.1 Literature Survey	1
1.2 Overview	2
1.3 Organization of The Project	3

CHAPTER 2: INTRODUCTION TO 5G COMMUNICATION

2.1 Introduction	4
2.2 Salient Features of 5G	5
2.2.1 Generation of 5G Communication	5
2.3 What is 5G Technology?	6
2.3.1 Timeline for every generation	7
2.3.2 Applications of 5G	8
2.4 Advantages of 5G Communications	8
2.4.1 Disadvantages of 5G Technology	8
2.4.2 Future Scope of 5G	9
2.4.3 Which countries are using 5G Technology?	10
CHAPTER 3: FILTER BANK MULTI-CARRIER	
3.1 Introduction to FBMC	11

3.1.1 Development of FBMC	11

3.1.2 What is FBMC	12
3.2 How data can travel in FBMC	14
3.3 Block diagram of FBMC	15
3.3.1 OQAM Pre-Processing	15
3.3.2 IFFT-Frequency Domain to Time Domain Conversion	15
3.3.3 Poly-Phase Network	15
3.3.4 Companding and De-Companding	17
3.3.5 Parallel to Serial Conversion	18
3.3.6 AWGN Channel	18
3.3.7 Serial to Parallel Conversion	19
3.3.8 FFT-Time Domain to Frequency Domain Conversion	19
3.3.9 OQAM Post Processing	20
3.4 Key Feature of FBMC	20
3.5 Orthogonal Frequency Division Multiplexing	20
3.5.1 Concept and Process	20
3.6 Difference between FBMC and OFDM	21
3.7 Advantages of FBMC system	23
3.7.1 Disadvantages of FBMC system	23
CHAPTER 4: NON-LINEAR COMPANDING TECHNIQUES	
4.1 Introduction to Non-Linear Companding	25
4.1.1 Applications of Companding	25
4.1.2 Working	25

4.1.3 Crest Factor	26
4.2 A-Law	26
4.3 Mu Law	27
4.4 Consequences of Companding	28
4.5 PAPR (Peak to Average Power Ratio)	29
CHAPTER 5: MATLAB	
5.1 Introduction to MATLAB	31
5.2 The MATLAB System	32
5.2.1 Development Environment	32
5.2.2 The MATLAB Algebraic Function	32
5.2.3 The MATLAB Language	32
5.2.4 Graphics	33
5.2.5 The MATLAB Appliance Affairs Interface (AP)	33
5.2.6 MATLAB Desktop	34
5.2.7 Using the MATLAB Editor to actualize M-Files	34
5.2.8 Getting Help	35
5.3 Communication	35
5.4 Key Features	36
5.5 System Design	36
5.6 System Characteristics	36
5.7 BER	37
5.8 Performance Visualization	37

5.9 Analog and Digital Modulation	37
5.10 Source and Channel Coding	38
5.10.1 Source Coding	38
5.10.2 Channel Coding	38
CHAPTER 6: SIMULATION RESULTS	
6.1 Power Spectral Density (PSD)	39
6.2 Peak to Average Power in FBMC System	40
6.2.1 CCDF vs PAPR (dB) plots in FBMC System with different	
companding techniques	41
Chapter 7: CONCLUSION AND FUTURE SCOPE	
7.1 Conclusion	45
7.2 Future Scope	45

REFERENCES

LIST OF FIGURES

- 1. Fig 2.1 Introduction to 5G Communication.
- 2. Fig 2.2 Features of 5G
- 3. Fig 2.3 Different Generation Techniques.
- 4. Fig 2.4 Graph-1 showing the timeline of all Previous Generation Techniques.
- 5. Fig 2.5 Future Scope in 5G.
- 6. Fig 3.1 The Graphical Illustration of the FBMC Transmitter.
- 7. Fig 3.2 The Graphical Illustration of the Generic FBMC Receiver.
- 8. Fig 3.3 OQAM Signaling.
- 9. Fig 3.4 Proposed Transceiver Structure of FBMC with Companding.
- 10. Fig 3.5 Frequency Response of FBMC and OFDM.
- 11. Fig 4.1 Compression curves for different values of compression parameters in(a) μ-law (b) A-law.
- 12. Fig 6.1 Power Spectral Density.
- 13. Fig 6.2 Conventional FBMC.
- 14. Fig 6.3 FBMC with Mu Law Companding.
- 15. Fig 6.4 FBMC with A Law Companding.
- 16. Fig 6.5 FBMC with Different Companding.

LIST OF TABLES

TABLE -1 PAPR in FBMC with Different Companding.

LIST OF ABBREVIATIONS

- 1. FBMC Filter Bank Multi Carrier
- 2. OFDM Orthogonal Frequency Division Multiplexing
- 3. ISI Inter Symbol Interference
- 4. ICI Inter Carrier Interference
- 5. CFO Carrier Frequency Offset
- 6. CP Cyclic Prefix
- 7. AWGN Add White Gaussian Noise
- 8. FFT Fast Fourier Transform
- 9. MUX Multiplexer
- 10. DEMUX Demultiplexer
- 11. PSD Power Spectral Density
- 12. 5G Fifth Generations
- 13. LTE Long Term Evolution
- 14. CP-OFDM Cyclic Prefix Orthogonal Frequency Division Multiplexing
- 15. BER Bit Error Rate
- 16. SNR Signal to Noise Ratio
- 17. PAPR Peak to Average Power Ratio
- 18. MIMO Multiple Input Multiple Output
- 19. OQAM Offset Quadrature Amplitude Modulation
- 20. ACE Active Constellation Extension
- 21. POCS Projection onto Convex Sets
- 22. SGP Smart Gradient Projection
- 23. IP Internet Protocol
- 24. Wi-Fi Wireless Fidelity
- 25. IFFT Inverse Fast Fourier Transform
- 26. dB Decibels
- 27. QPSK Quadrature Phase Shift Keying
- 28. S/P Serial to Parallel Converter
- 29. AFB Analysis Filter Bank

- 30. SFB Synthesis Filter Bank
- 31. CFO Carrier Frequency Offset

CHAPTER 1

INTRODUCTION

1.1 Literature Survey

Filter Bank Multicarrier is a form of multicarrier modulation technique that is highly used for high data rate cellular and wireless systems. We heard this name first in a paper "The potential of FBMC over OFDM for the Future 5G Mobile Communication Technology", by A.N. Ibrahim, M.F.L. Abdullahi, who is the faculty of electrical and electronics engineering University Hussein Onn Malaysia, that FBMC is a combination of multiplexing and modulation with function by breaking the wide band channel into a number of narrow band channels which is called sub-channels. The complicated modulation values of FBMC systems will be spread over several carriers and filtered by a prototype filter. The simple concept and low complexity in OFDM systems makes the FBMC systems received a limited attenuation but the FBMC systems offers more robustness to the time and frequency offset than OFDM and does not use any cyclic prefix extension. In FBMC system, the signal with high spectral containment will be used to reduce the sidelobes of each subcarrier frequency. FBMC gives better bandwidth efficiency than OFDM from the paper "Improvement of FBMC over OFDM in terms of PSD and BER", by A. Lakshmi Narayana, faculty of electronics and communication engineering, Anil Neerukonda Institute of Technology and Sciences, Andhra Pradesh, says that the power spectral density is more in FBMC system comparatively than OFDM system. Also, out of FBMC and OFDM, FBMC has high PAPR. From the paper "The potential of FBMC over OFDM for the Future 5G Mobile Communication Technology", by A.N. Ibrahim, M.F.L. Abdullahi, who were the faculty of electrical and electronics engineering, University Tun Hussein Onn Malaysia, that SNR is more in FBMC than OFDM and BER is less in FBMC than OFDM. From the paper, "Performance Evaluation of PAPR Reduction in FBMC System using Non-Linear Companding Transform", by Imad A Shebeen, Abdelhalim Zekry, Faculty of Electronics and Communication Engineering, Ain Shams University, Cairo-Egypt, that the main drawback of FBMC system is its high Peak to

Average Power Ratio (PAPR) of the transmitted signal. PAPR occurs when in a multicarrier system, the different sub-carriers are out of phase with each other. All such instant they are different with respect to each other at different phase values. When all the points achieve the maximum value simultaneously, this will cause the output envelope. Due to the presence of large number of independently modulated sub-carriers FBMC system the peak value of the system can be very high as compared to the average of the whole system. This high PAPR reduces the efficiency of the power amplifier. We got to know that Non-Linear companding is one of the best PAPR reduction techniques used in multicarrier modulation due to less complexity and good BER (Bit Error Rate).

1.2 Overview:

The demand for high data services is gained more importance in present generation. As we all know, data transmission is of 2 types i.e., wireless and wired medium. These data services must be very reliable for transmission of data even in harsh environment. Most of this transmission systems experience degradation such as noise, large attenuation, time variance, multipath non-linearities. Multi-carrier modulation technique is a physical layer technique that has gained a lot of importance in present days due to its robustness. In multi-carrier modulation, the most commonly used technique is OFDM (Orthogonal Frequency Division Multiplexing), FBMC (Filter Bank Multicarrier), which has recently become very popular in wireless communication. There is a great demand of advance communication and according to it there should be high data rate in addition to both efficiency and low bit error rate.

Filter Bank Multicarrier with Offset Quadrature Amplitude Modulation System (FBMC/OQAM) as one of the many new waveforms is proposed for the next wireless communication system (5G). The FBMC system is a multicarrier modulation scheme with relaxed orthogonally, increased frequency efficiency, improved shape and low out-of-band interference inherent in FBMC/OQAM techniques. In OFDM, entire band is separated, but in FBMC, each sub-carrier is shifted independently. When carriers are modulated using FBMC, side lobes are removed and gives clear outcomes.

The main drawback of FBMC/OQAM system is a high Peak to Average Power Ratio (PAPR) of the transmitted signal. To overcome that, many researchers proposed schemes to reduce PAPR for FBMC system. Active Constellation Extensions (ACE) with projection onto convex sets (POCS) and smart gradient projection (SGP), are used for reducing PAPR of FBMC system. But they increase the average power transmitted leading to a decrease in the power efficiency and degrading the bit error rate (BER) performance as well as they require more implementation complexity.

Nonlinear Companding is one of the best PAPR reduction techniques used in multicarrier modulation due to less complexity and good bit error rate BER. The companding function is used in order to attenuate the high peaks and amplify the low amplitudes. At the receiver, the inverse companding function is applied in order to recover the original signal. Moreover, by choosing proper companding parameters, the average transmitted power can be kept unchanged after companding. In this paper, we used different nonlinear companding techniques to reduce PAPR of FBMC system. The companding techniques used are Mu-law and A-law companding.

1.3 Organization of the Project:

The project report is organised in six chapters. It starts from the introduction. The current Chapter 1 introduces the project and gives a brief description of Filter Bank Multi Carrier compared to Orthogonal Frequency Division Multiplexing technique. Chapter 2 provides the brief idea about the 5G Communication and it gives the introduction to the 5G. Chapter 3 gives the detailed working of the Filter Bank Multi Carrier and detailed study of the block diagram. Chapter 4 talks about the Non-Linear companding techniques i.e., about A law, Mu law, and also about PAPR(Peak to Average Power Ratio). Chapter 5 talks about the introduction to MATLAB. Chapter 6 gives the Result, Conclusion and Future scope of 5G technology.

CHAPTER 2

INTRODUCTION TO 5G COMMUNICATION

2.1 Introduction:

5G is the Fifth-Generation technology. It has many advanced features potential enough to solve many of the problems of our mundane life. It is beneficial for the government, as it can make governance easier; for the students, as it can make available the advanced courses, classes, and materials online; it is easier for the common people as well, as it can facilitate them on the internet everywhere. So, this tutorial is divided into various chapters and describes the 5G technology, its applications, challenges, etc.

Radio technologies have evidenced a rapid and multidirectional evolution with the launch of the analogue cellular systems in the 1980s. Thereafter, digital wireless communication systems are consistently on a mission to fulfil the growing need of human beings (1G, \dots 4G, or now 5G).



Fig 2.1 Introduction to 5G Communication

So, this article describes the 5G technology emphasizing on its salient features, technological design (architecture), advantages, shortcomings, challenges, and future scope.

2.2 Salient features of 5G:

5th Generation Mobile Network or simply 5G is the forthcoming revolution of mobile technology. The features and its usability are much beyond the expectation of a normal human being. With its ultra-high speed, it is potential enough to change the meaning of a cell phone usability.

With a huge array of innovative features, now your smartphone would be more parallel to the laptop. You can use broadband internet connection; other significant features that fascinate people are more gaming options, wider multimedia options, connectivity everywhere, zero latency, faster response time, and high-quality sound.

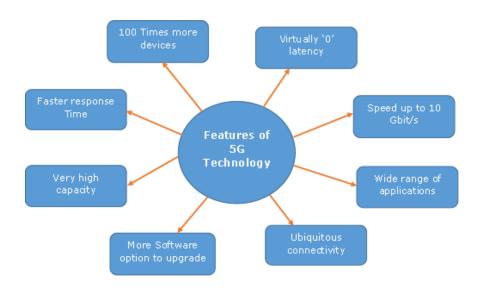


Fig 2.2 Features of 5G

2.2.1 Generations of 5G Communications:

If we look back, we will find that every next decade, one generation is advancing in the field of mobile technology. Starting from the First Generation (1G) in 1980s, Second Generation (2G) in 1990s, Third Generation (3G) in 2000s, Fourth Generation (4G) in 2010s, and now Fifth Generation (5G), we are advancing towards more and more sophisticated and smarter technology.



Fig 2.3 Different Generation Technologies

2.3 What is 5G Technology?

The 5G technology is expected to provide a new (much wider than the previous one) frequency bands along with the wider spectral bandwidth per frequency channel. As of now, the predecessors (generations) mobile technologies have evidenced substantial increase in peak bitrate. Then — how is 5G different from the previous one (especially 4G)? The answer is — it is not only the increase in bitrate made 5G distinct from the 4G, but rather 5G is also advanced in terms of –

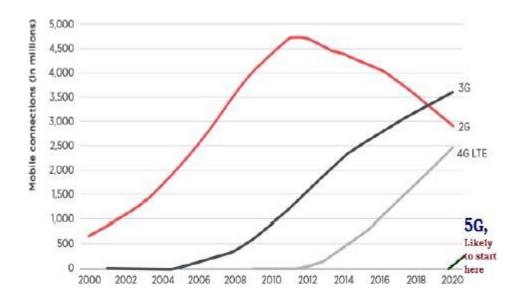
- High increased peak bit rate
- Larger data volume per unit area (i.e., high system spectral efficiency)
- High capacity to allow more devices connectivity concurrently and instantaneously
- Lower battery consumption
- Better connectivity irrespective of the geographic region, in which you are
- Larger number of supporting devices

- Lower cost of infrastructural development
- Higher reliability of the communications

As researchers say, with the wide range of bandwidth radio channels, it is able to support the speed up to 10 Gbps, the 5G Wi-Fi technology will offer contiguous and consistent coverage-wider area mobility in true sense.

2.3.1 Timeline for every generations

Normally, it is expected that the time period required for the 5G technology development and its implementation is about five years more from now (by 2020). But to become usable for the common people in developing countries, it could be even more.





By considering the multiple utility and various fashionable salient features, researchers are anticipating that this technology will be in use until the 2040s.

2.3.2 Applications of 5G

It will make a unified global standard for all.

- Network availability will be everywhere and will facilitate people to use their computer and such kind of mobile devices anywhere anytime.
- Because of the IPv6 technology, visiting care of mobile IP addresses will be assigned as per the connected network and geographical position.
- Its application will make the world a real Wi-Fi zone.
- Its cognitive radio technology will facilitate different versions of radio technologies to share the same spectrum efficiently.
- Its application will facilitate people to avail radio signal at higher altitude as well.

2.4 Advantages of 5G Communications

- High resolution and bi-directional large bandwidth shaping.
- Technology to gather all networks on one platform.
- More effective and efficient.
- Technology to facilitate subscriber supervision tools for the quick action.
- Most likely, will provide a huge broadcasting data (in Gigabit), which will support more than 60,000 connections.
- Easily manageable with the previous generations.
- Technological sound to support heterogeneous services (including private network).
- Possible to provide uniform, uninterrupted, and consistent connectivity across the world.

2.4.1 Disadvantages of 5G Technology

- Technology is still under process and research on its viability is going on.
- The speed, this technology is claiming seems difficult to achieve (in future, it might be) because of the incompetent technological support in most parts of the world.
- Many of the old devices would not be competent to 5G, hence, all of them

need to be replaced with new one — expensive deals.

- Developing infrastructure needs high cost.
- Security and privacy issues yet to be solved.

2.4.2 Future Scope of 5G

Several researches and discussions are going on across the world among technologists, researchers, academicians, vendors, operators, and governments about the innovations, implementation, viability, and security concerns of 5G.

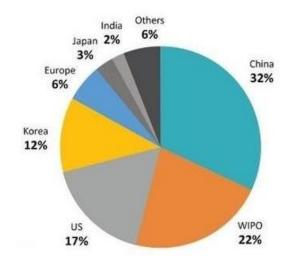
As proposed, loaded with multiple advanced features starting from the super high speed internet service to smooth ubiquitous service, 5G will unlock many of the problems. However, the question is — in a situation, where the previous technologies (4G and 3G) are still under process and in many parts yet to be started; what will be the future of 5G?



Fig 2.5 future scope in 5G

5th generation technology is designed to provide incredible and remarkable data capabilities, unhindered call volumes, and immeasurable data broadcast within the latest mobile operating system. Hence, it is more intelligent technology, which will interconnect the entire world without limits.

Likewise, our world would have universal and uninterrupted access to information, communication, and entertainment that will open a new dimension to our lives and will change our lifestyle meaningfully.



2.4.3 Which countries are using 5G Technology?

Fig 2.6 Graphical distribution of 5G Patent Families

First adopters embracing 5G will gain a competitive edge over the others, and it looks like China is currently leading the way. According to various sources, Asian countries like China, South Korea, Japan, and the US (outside Asia) are investing and building 5G infrastructure.

Since 2015, China has spent the most money on 5G and the amount is about US\$24 billion which is more than any other countries. These numbers are momentous, as the potential economic impact of 5G far outweighs the improvements recorded when the industry transitioned from 3G to 4G.

CHAPTER 3

FILTER BANK MULTI-CARRIER

3.1 Introduction to FBMC

Filter Bank Multicarrier, FBMC is a form of multicarrier modulation in which the carriers are filtered to provide a more spectral efficient form of waveform. Filter Bank Multicarrier, FBMC is a form of multi-carrier modulation that has its origins within OFDM. It is a development of OFDM and aims to overcome some of the issues, although this comes at the cost of increased signal processing.

FBMC has a much better usage of the available channel capacity and is able to offer higher data rates within a given radio spectrum bandwidth, i.e., it has a higher level of spectrum efficiency. Filter bank multicarrier aims to overcome some of the shortcomings that were encountered with OFDM, orthogonal frequency division multiplexing. One of the main shortcomings arises from the fact that OFDM requires the use of what is termed as cyclic prefix. The cyclic prefix is essentially a copy part of a transmitted symbol in OFDM that is appended at the beginning of the next which is considered to be a major drawback and it is eliminated using FBMC. The reason behind the wide usage of the FBMC system are it is a multicarrier modulation scheme with relaxed orthogonality, it has increased frequency efficiency, improved shape, and low out-of-band interference.

3.1.1 Development of FBMC

The use of FBMC and multicarrier modulation in general has come to the fore in recent years as it provides an ideal platform for wireless data communications transmissions.

However, the concept of FBMC technology was first investigated in the mid-1960s during research into methods for reducing interference between closely spaced channels and also avoid the use of cyclic prefix which results in increased complexity. In addition to these other requirements needed to achieve error free data transmission in the presence of interference and selective propagation conditions.

Initially the use of FBMC mainly aims at removing cyclic prefix that is used in OFDM but later on it is found that there are some drawbacks such as increased Peak to Average Power Ratio which resulted in power amplifier inefficiency but later on many schemes have been proposed to reduce PAPR which resulted in high efficiency enabling FBMC to be considered for the 5G mobile communications systems which started to be deployed from around 2019. Also, FBMC was adopted for Wi-Fi and a variety of other wireless data systems.

3.1.2 What is FBMC

FBMC systems are a subclass of Multicarrier (MC) systems. FBMC modulation is a multicarrier modulation method in which a set of synthesis and analysis filters are employed at the transmitter and receiver respectively. The filters use in the FBMC systems are a set of bandpass filters. In this filter, the frequency shifted or modulated versions of a prototype lowpass filter. FBMC offers a better spectral containment than OFDM as the filter bandwidth, so the selectivity is a parameter that can be assorted during lowpass prototype design. Besides that, FBMC gives the better bandwidth efficiency compared to OFDM. This is because, FBMC does not use the CP extension, so it will be attenuated the interferences within and close to the used frequency band efficiently. Next, the FBMC systems are comparatively more resistant to narrowband noise effects.

Figure 3.1 shows the graphical illustration of the FBMC transmitter meanwhile Fig 3.2 shows the graphical illustration of a generic FBMC receiver. At the transmitter as shown in Fig. 3.1, the high-speed input signal will be demultiplexed into N branches. After that, it will be modulated by the different or same signal constellation as required. The subsequent modulated branches are the unsampled to give N copies. The unsampled data will be sent through the set of synthesis filters gk(n), k=0,1,...,N-1. Next, to produce the transmitted signal s(n), the output of all filters will be summed together. At the receiver as shown in Fig. 3.2, to give N subcarriers of different center frequencies, the received signal r(n) will be passed through to the bank of analysis filters fk(n),k=0,1,...,N-1. The signal in every branch will be down

sampled by N, demodulated and multiplexed to produce the estimate of the original signal Xr(n).

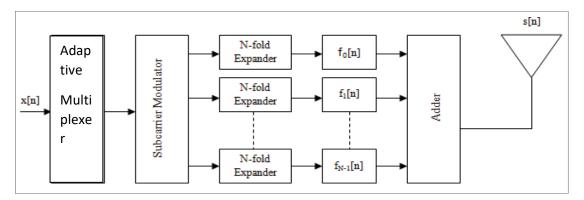


Fig 3.1 The graphical illustration of the FBMC transmitter

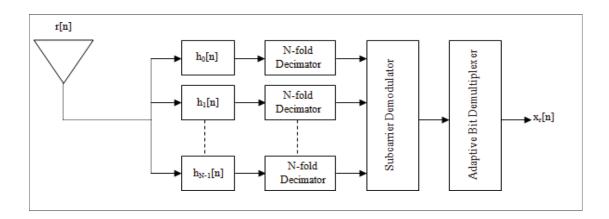


Fig 3.2 The graphical illustration of a generic FBMC receiver

3.2 How data can travel in FBMC

The traditional format for sending data over a radio channel is to send it serially, one bit after another. This relies on a single channel and any interference on that single frequency can disrupt the whole transmission.

FBMC adopts a different approach. First the data is transmitted in parallel across the various subcarriers and then there is a shift of half the inverse of the sub-channel between the real part and the imaginary part of a complex symbol such that there is staggering of in-phase and quadrature-phase components in both time and frequency domains, as shown in Fig. 9 and hence, orthogonality is maintained within the real and imaginary domains separately.

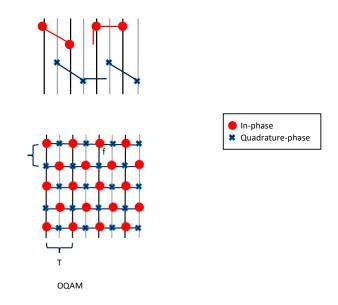
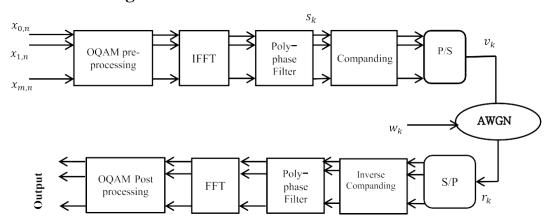


Fig 3.3 OQAM signalling.

The distribution of data with an offset introduced between the real and imaginary parts has some further advantages. Due to the presence of offset the inter symbol interference and inter carrier interference can be eliminated to some extent. By performing proper post processing at the receiver enables us to reconstruct the data.



3.3 Block Diagram of FBMC



3.3.1 OQAM Pre-Processing

The function of OQAM pre-processing is to take the serial to parallel converted data that is obtained from the random generator and then perform the quadrature amplitude modulation which results in complex data, and then there is a shift of half the inverse of the sub-channel between the real part and the imaginary part of a complex symbol such that there is staggering of in-phase and quadrature-phase components in both time and frequency domains the output of the OQAM pre-processing is then passed through the IFFT block.

3.3.2 IFFT-Frequency-Domain to Time Domain Conversion

The orthogonality of the subcarrier is maintained and the frequency domain signals are converted into a time domain and the generation of real-output signal is achieved by arranging the conjugate of sub-carrier. In this stage the techniques like IFFT Mapping, zero mapping and selector bank are included to overcome the problem of length of subcarrier and bin size.

3.3.3 Polyphase Network

In order to fix the size of FFT some additional processing is required, called as polyphase network. The prototype filter is defined as a set of coefficients and the relationship between input and output sequences, which is

$$\mathbf{y}(\mathbf{n}) = \sum_{i=0}^{L-1} h_i \mathbf{x}(n-i)$$

The filter impulse response, of length L, is the sequence of coefficients hh_i the frequency response is expressed by

$$H(f) = \sum_{i=0}^{L-1} h_i e^{-j_{2\pi} i f}$$

The filter has linear phase if the coefficients are symmetrical and, in this case, the delay is

$$\tau = \frac{L-1}{2}$$

In digital signal processing, and particularly in digital filtering, it is customary to use the Z- transfer function, which generalizes the frequency response and is defined by

$$H(Z) = \sum_{i=0}^{L-1} h_i z^{-i}$$

The filter frequency response is the restriction of the Z-transfer function to the unit circle, i.e., it is obtained by letting $Z = e^{j^2 \pi f}$.

Now, if we assume, as in the previous section, that the filter length is a product of two factors,

L = K.M, the sequence of filter coefficients can be decomposed into M interleaved sequences of K coefficients and the Z-transfer function can be expressed as a double summation

It turns out that each individual filter element has a frequency response of a phase shifter hence the name of polyphase decomposition and polyphase network for the complete set.

Now, turning to the filter bank in the transmitter, which is generated by shifting the

response of the prototype filter on the frequency axis, a global Z-transfer function can be derived Shifting the frequency response of the filter function H (f) by 1/M on the frequency axis leads to the function

B(f) = H(f-1/M)

The key point here is that the functions H (Z^M) are not affected by the frequency shift. Then, considering all the shifts by multiples of 1/M and the associated filters, and letting W = $e^{-j^2\pi/M}$, a matrix equation is obtained which is the inverse discrete Fourier transform and all the filters in the bank have the same filter elements.

In the implementation, the transmitter output is the sum of the outputs of the filters of the bank. Thus, the processing associated with the filter elements H (Z^M) can be carried out after the summation which is performed by the iDFT.

The same scheme applies to the filter bank in the receiver. The difference is that the frequency shifts are multiples of -1/M and the discrete Fourier transform (DFT) replaces the iFFT. In fact, for each sub-channel, the signal of interest is shifted around the frequency origin and filtered. Again, the filter elements are the same for all the filters in the bank and, since it is the sum of the sub-channel signals which is received, the processing can be common and the separation of the signals can take place afterwards, with the help of the DFT.

3.3.4 Companding and De-Companding

The output of the polyphase network will result in high PAPR(Peak to Average Ratio) at the transmitter end which results in increased battery consumption and decreases power amplifier efficiency, for example if PAPR is 10dB and if we want to pass a signal with power 0.2W then the power amplifier must be able to handle a peak power 2W in order to reduce the PAPR we use companding techniques which uses a companding function and if the input is less than the amplification of the input takes place whereas if the input if small then the attenuation takes place. At the receiver the reverse process takes place by using De companding function in order to reproduce the original data.

3.3.5 Parallel to Serial Conversion

The final stage in the implementation must undo the first stage. A switch is used to time- division multiplex the four individual bit signals into a single sequence.

3.3.6 AWGN Channel

Additive white Gaussian Noise (AWGN) is a basic noise model used in Information theory to imitate the effect of many random processes that occur in nature.

The term additive white Gaussian noise (AWGN) originates due to the following reasons:

Additive: The noise is additive, i.e., the received signal is equal to the transmitted signal plus noise. This gives the most widely used equality in communication systems which is shown in Figure below.

r(t)=s(t)+w(t)(1)(1)r(t)=s(t)+w(t)

Moreover, this noise is statistically independent of the signal. Remember that the above equation is highly simplified due to neglecting every single imperfection a Tx signal encounters, except the noise itself.

White: Just like the white colour which is composed of all frequencies in the visible spectrum, white noise refers to the idea that it has uniform power across the whole frequency band. As a consequence, the Power Spectral Density (PSD) of white noise is constant for all frequencies ranging from $-\infty-\infty$ to $+\infty+\infty$, as shown in Figure below.

Nyquist investigated the properties of thermal noise and showed that its power spectral density is equal to $k \times Tk \times T$, where kk is a constant and TT is the temperature in Kelvin. As a consequence, the noise power is directly proportional to the equivalent temperature at the receiver.

Gaussian: The probability distribution of the noise samples is Gaussian with a zero mean, i.e., in time domain, the samples can acquire both positive and negative values and in addition, the values close to zero have a higher chance of occurrence while the

values far away from zero are less likely to appear. This is shown in Figure below. As a result, the time domain average of a large number of noise samples is equal to zero.

In reality, the ideal flat spectrum from $-\infty-\infty$ to $+\infty+\infty$ is true for the frequencies of interest in wireless communications (a few kHz to hundreds of GHz) but not for higher frequencies. Nevertheless, every wireless communication system involves filtering that removes most of the noise energy outside the spectral band occupied by our desired signal. Consequently filtering, it is not possible to distinguish whether spectrum was ideally flat

3.3.7 Serial to Parallel Conversion

The main function of serial to parallel converter is to convert the serial data parallelly. The parallel data is transmitted by assigning a unique word to each of the subcarriers. Once the symbol has been allocated to each of the subcarriers then they are phased mapped according to the modulating scheme. Digital Modulation SchemeUse in FBMC Transmitter

It is one of the advantages of FBMC that different modulation schemes can be applied to each sub channel depending on channel condition, data rate, robustness, throughput and channel bandwidth. There could be different modulation schemes applied specified by complex numbers i.e., QPSK, 16 QAM. Modulation to FBMC sub channels can be made adaptive after getting information and estimation of channel at transmitter.

3.3.8 FFT: Time Domain to Frequency Domain Conversion

FBMC distributes the data over a large number of carriers at different frequencies. This spacing provides the orthogonality which prevents the receivers from seeing the wrong frequencies. In contrast to other multi-carriers' techniques, like OFDM, FBMC prevents the Inter Symbol Interference (ISI) by using a poly-phase network is the key features of FBMC is the IFFT/FFT pair. These two mathematical tools are used to transform several signals on different carriers from the frequency domain to the time-domain in the IFFT (or FFT-1) and from the time-domain to the frequency-domain in the FFT.

3.3.9 OQAM Post-Processing

The main function of OQAM post-processing is to separate the alternate real and imaginary parts which are separated by an offset when passed through the transmitter and then demodulation process takes place which results in the desired data at the receiver end.

3.4 Key Features of FBMC

- In the FBMC system, the signal with high spectral containment will be used to reduce the sidelobes of each subcarrier frequency.
- The subcarriers are orthogonal to each other.
- The FBMC system offers more robustness to the time and frequency offset.
- The FBMC system offers more frequency spectral efficiency

3.5 Orthogonal Frequency Division Multiplexing

In telecommunications, orthogonal frequency-division multiplexing is a type of digital transmission and a method of encoding digital data on multiple carrier frequencies.

3.5.1 Concept and Process

Orthogonal Frequency Division Multiplexing (OFDM) is one of the modulation types used for current wireless and telecommunications systems. This system used the technique of encoding digital data on multiple carrier frequency and becomes a popular method for wideband digital communication. It is widely used to produce high data rates and combating multipath fading in wireless communication technology. OFDM is already used over the world to attain high data rates which is needed for data intensive applications. It has been used in wireless network, audio broadcasting and 4G mobile communication technology. This modulation format already been used in the WIFI arena (802.11a, 802.11ac, etc). OFDM use the Cyclic Prefix (CP) which will reduce the overall spectral efficiency. OFDM based on the idea of modulating each data stream on subcarriers and dividing high-bit-rate data

stream into several lower bit-rate data. Conventional OFDM makes use of Fast Fourier Transform (FFT) as its basic block. Multicarrier modulation knowing as schemes which able to provide high data rate. Figure 4 shows the frequency response for OFDM which exhibits strong sidelobes due to rectangular windowing. OFDM is a wideband modulation technique which is able to handle with the issues of the multipath reception by transmitting many narrowband overlapping digital signals in parallel in one wide band. It is very useful for communication over channels with frequency selective fading. Nevertheless, it is difficult in handling selective fading in the receiver because of the complicate architecture of the receiver. Besides that, flat fading is easy to combat compared to the frequency selective fading by the use of simple error correction and equalization schemes.

3.6 Differences between FBMC and OFDM

OFDM and FBMC known as the multicarrier techniques which is the data symbols will be transmitted simultaneously over the multiple frequency subcarriers. Their nature of multicarrier signals gives in-build support for frequency selective link/rank adoption. The main difference between OFDM and FBMC is the pulse shaping applied at each subcarrier. Most of the wireless mobile communication technology that was developed today is based on OFDM. FBMC is an advancement of OFDM. The basic change in the FBMC system is the replacement of the CP in OFDM with the multicarrier system based on filter bank [24], the inverse Fast Fourier transform (IFFT) and the input of the Cyclic Prefix was replaced by the synthesis filter bank (SFB). Meanwhile for the output of the Cyclic Prefix and Fast Fourier transform (FFT) was replaced by the analysis filter bank (AFB). The frequency spectrums of the subcarriers in the OFDM system are overlapped with the minimum frequency spacing. Besides that, the orthogonality was reached between the different of the subcarriers. The input stream was split into parallel data streams using the serial to parallel data converter (S/P). Subsequently, to generate time sequence of the streams, it will be passed into an IFFT block. The symbol time sequences of the OFDM system will be extended by adding CP. The digital signal produced and converted into analog form before it has been transmitted over the channel. FBMC overcomes the weakness of the OFDM system. In the FBMC system, it produced a well localized subchannel in both frequency and time domain by adding a generalized pulse shaping filters. There is no CP needed in the FBMC system, so it provided more effective use of the radio resources and have more spectral containment signals. Filter bank can be defined as an array of the N filters that will be processed the N input signals to generate N outputs. There are two types of the filter bank in this system which is analysis filter bank (AFB) and synthesis filter bank (SFB). In the AFB, the input of N filter is connected together. Besides that, the system in analogous manner that can be assumed as an analyzer to the input signal based on each filter characteristics. Meanwhile in the SFB, the outputs of the filter array will be added and a new signal will be synthesized. The synthesized analysis configuration known as trans multiplexer and will be applied in the multicarrier communication systems. Figure 9 shows the frequency response for OFDM and FBMC. Based on this figure, it shows the major impact to the transmission data. The energy was concentrated within the frequency range of a single subcarrier for the FBMC systems, meanwhile in the OFDM system shows the rather strong sidelobes due to rectangular windowing. Hence we conclude that FBMC is relatively more efficient when compared with OFDM and it has major advantages which overcome the drawback of OFDM system and it is quite useful for next generation technology such as 5G and many more, the reason behind its success is efficient use of data and complexity is less when compared with other multiplexing techniques.

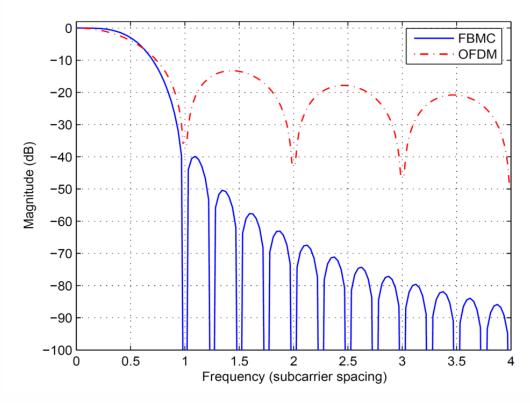


Fig 3.5 Frequency Response of FBMC and OFDM

In the OFDM system, CP extension required, so it reduces the bandwidth efficiency, but in the FBMC system, CP are not required and hence conserves the bandwidth. Besides that, OFDM is very sensitive to the carrier frequency offset (CFO). Meanwhile for the FBMC system, it less sensitive, so it performs significantly with the increase of the consumer mobile network, it shows that the FBMC system improves in performance compared to OFDM as shown in Fig.3.3.

3.7 Advantages of FBMC system

- FBMC is able to provide a spectrum efficient and more selective system.
- The cyclic prefix, CP required for OFDM is not needed thereby freeing up more space for real data.
- The FBMC system offers more robustness to the time and frequency offset.
- Provide robust narrowband jammers.

3.7.1 Disadvantages of FBMC System

• The use of MIMP with FBMC is very complicated and as a result few systems

have investigated the use of these two techniques together.

- The design of wide bandwidth and high dynamic range systems with FBMC provides some significant RF development challenges.
- FBMC is more complicated than OFDM it introduces an overhead in overlapping symbols in the filter bank in the time domain.

CHAPTER 4

NON-LINEAR COMPANDING TECHNIQUES

4.1 Introduction to Non-Linear Companding

Nonlinear Companding is a special type of clipping scheme used to provide great PAPR reduction with better BER performance and less implementation complexity required to build the system. The companding transforms enlarge the small signals while compressing the large signals to increase the immunity of small signals from noise and interference. The companding function is applied at the transmitter to compand the output signals using strict monotone function and at the receiver, the inverse companding function is used to recover the signal in the receiver. Different nonlinear companding techniques have been discussed below.

4.1.1 Applications of Companding

Companding is used in digital telephony systems, compressing before input to an analog-to-digital converter, and then expanding after a digital-to-analog converter. This is equivalent to using a non-linear ADC as in a T-carrier telephone system that implements A-law or μ -law companding. This method is also used in digital file formats for better signal-to-noise ratio (SNR) at lower bit depths. For example, a linearly encoded 16-bit PCM signal can be converted to an 8-bit WAV or AU file while maintaining a decent SNR by compressing before the transition to 8-bit and expanding after conversion back to 16-bit. This is effectively a form of lossy audio data compression.

Professional wireless microphones do this since the dynamic range of the microphone audio signal itself is larger than the dynamic range provided by radio transmission. Companding also reduces the noise and crosstalk levels at the receiver.

Companders are used in concert audio systems and in some noise reduction schemes.

4.1.2 Working

The electronic circuit that does this is called a compander and works by compressing or expanding the dynamic range of an analog electronic signal such as sound recorded by a microphone. One variety is a triplet of amplifiers: a logarithmic amplifier, followed by a variable-gain linear amplifier and an exponential amplifier. Such a triplet has the property that its output voltage is proportional to the input voltage raised to an adjustable power.

Companded quantization is the combination of three functional building blocks – namely, a (continuous-domain) signal dynamic range compressor, a limited-range uniform quantizer, and a (continuous-domain) signal dynamic range expander that inverts the compressor function. This type of quantization is frequently used in telephony systems.

In practice, companders are designed to operate according to relatively simple dynamic range compressor functions that are designed to be suitable for implementation using simple analog electronic circuits. The two most popular compander functions used for telecommunications are the A-law and μ -law functions.

4.1.3 Crest Factor

Crest factor is a parameter of a waveform, such as alternating current or sound, showing the ratio of peak values to the effective value. In other words, crest factor indicates how extreme the peaks are in a waveform. Crest factor 1 indicates no peaks, such as direct current or a square wave. Higher crest factors indicate peaks, for example sound waves tend to have high crest factors.

Crest factor is the peak amplitude of the waveform divided by the RMS value of the waveform

The peak-to-average power ratio (PAPR) is the peak amplitude squared (giving the peak power) divided by the RMS value squared (giving the average power). It is the square of the crest factor:

When expressed in decibels, crest factor and PAPR are equivalent, due to the way decibels are calculated for power ratios vs amplitude ratios.

Crest factor and PAPR are therefore dimensionless quantities. While the crest factor is defined as a positive real number, in commercial products it is also commonly stated

as the ratio of two whole numbers, e.g., 2:1. The PAPR is most used in signal processing applications. As it is a power ratio, it is normally expressed in decibels (dB). The crest factor of the test signal is a fairly important issue in loudspeaker testing standards; in this context it is usually expressed in dB.

The minimum possible crest factor is 1, 1:1 or 0 dB.

4.2 A-LAW

We know that in the companding technique, the compression of OFDM signals at the transmitter and expansion at the receiver. In this companding method, the compressor characteristic is piecewise, made up of a linear segment for low level inputs and a logarithmic segment for high level inputs. A-law compressor characteristics for different values of A. Corresponding to A=1, we observe that the characteristic is linear (no compression) which corresponds to a uniform quantization. A-law has mid riser at the origin. Hence it contains non-zero value. The practically used value of "A" is 87.6. The A-law companding is used for PCM telephone systems.

The linear segment of the characteristic is for low level inputs whereas the logarithmic segment is for high level inputs. It is mathematically expressed as,

The reason for this encoding is that the wide dynamic range of speech does not lend itself well to efficient linear digital encoding. A-law encoding effectively reduces the dynamic range of the signal, thereby increasing the coding efficiency and resulting in a signal-to-distortion ratio that is superior to that obtained by linear encoding for a given number of bits.

The A -LAW companding function which is applied to the end of the transmitter is given by

 $h(s_k) = u(A) \operatorname{sgn}(s_k) \{ A | s_k | / 1 + \ln(A), \quad \text{if } | s_k | < | s_k | \max / A$ $h(s_k) = u(A) \operatorname{sgn}(s_k) \{ 1 + \ln(A | s_k |) / 1 + \ln(|A|), \quad \text{if } | s_k | > = | s_k | \max / A$

Where the A value tells us about the order of companding level at the transmitter side, where k(A) is normalization constant. To recover the original signal, the De companding function is used at the receiver side which is given by

$$h(r_{k}) = k (A)^{(-1)} sgn(r_{k}) \{ |r_{k}|(1+\ln(A))/A \quad \text{if, } |r_{k}| < |r_{k}| max/1 + \ln(A) \}$$

$$h(r_{k}) = k (A)^{(-1)} sgn(r_{k}) \{ exp (|r_{k}|(1+\ln(A)))/A \quad \text{if, } |r_{k}| > |r_{k}| max/1 + \ln(A) \}$$

Where k(A) is normalization constant.

4.3 *μ***-Law**

In the μ -law companding, the compressor characteristic is piecewise, made up of a linear segment for low level inputs and a logarithmic segment for high level inputs. μ -law compressor characteristics for different values of μ . Higher the value of μ , more is the compression. Corresponding to μ =0, we observe that the characteristic is linear (no compression) which corresponds to a uniform quantization. μ -law has mid tread at the origin. Hence it contains zero value. The practically used value of " μ " is 255. The μ -law companding is used for speech & music signals. This μ -law companding technique is used in United States (U.S.), Canada, Japan, etc. The early BELL digital transmission system uses a 7-bit PCM Coder with μ =100. But the most recent digital transmission system uses an 8-bit PCM code with the value of μ =255. Where the Mu ratio is used to control the order of companding level at the transmitter side, where u(Mu) is normalization constant.

The MU - LAW companding function which is applied to the end of the transmitter is given by

 $h(s_k) = u(Mu) \operatorname{sgn}(s_k) \ln (1+Mu|s_k|)/\ln(1+Mu)$

Where the Mu ratio tells us about the order of companding level at the transmitter side, where u(Mu) is normalization constant. To recover the original signal, the De companding function is used at the receiver side which is given by

 $h(r_k) = u(Mu)^{(-1)} sgn(r_k)(1/Mu)((1+Mu)^{|r_k|} - 1)$

4.4 Consequences of Companding

In companding, the rate of compression depends on the values of compression parameters μ and A of the equations presented in the section "Logarithmic Companding Curves". Greater the value of compression parameters, higher is the rate of compression (Figure 4.1).

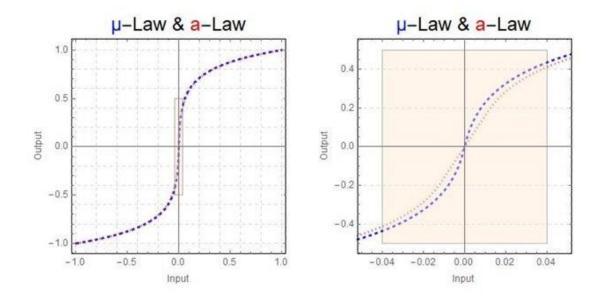


Fig 4.1 Compression curves for different values of compression parameters in (a) μ-law (b) A-law.

Higher compression rates imply greater non-linearity in quantization, which means there is a better representation of lower amplitude signals (using more number of bits) compared to higher amplitude ones. This indicates that, in a companded signal, the quantization error will be at its minimum at low levels and will gradually increase with an increase in the level of the input signal (PDF). In addition, the smaller the quantization interval, the better the signal-to-quantization noise ratio (SQNR). This means companding increases the SQNR at low-level signals while degrading it for higher amplitude ones (PDF).

The scenario well suits the demand for telephone systems which primarily transmit human speech wherein low amplitude quieter phonemes occur more frequently when compared to high amplitude louder phonemes (PDF). A direct consequence of this is an improvement in the quality of the audible signal, as we would have accounted for the sensitivity issues posed by the human ear.

4.4 PAPR (Peak to Average Power Ratio)

The main issue in FBMC system is high PAPR which reduces the efficiency of the power amplifier. The FBMC system is used to send complex symbols which modulate different subcarriers resulting in a high PAPR. PAPR of an FBMC system is defined as the ratio of peak power to the average power [11]. In general, the PAPR can be written as

$$PAPR_{dB} = 10 \log \left(\frac{max[x(t)x^*(t)]}{E[x(t)x^*(t)]} \right)$$

The linear power amplifiers are being used in the transmitter so the Q-point must be in the linear region. Due to the high PAPR the Q-point moves to the saturation region hence the clipping of signal peaks takes place which generates in-band and out-ofband distortion. So, to keep the Q-point in the linear region the dynamic range of the power amplifier should be increased which again reduces its efficiency and enhances the cost. Hence a trade-off exists between nonlinearity and efficiency. And also, with the increasing of this dynamic range the cost of power amplifier increases. As a communication engineer our objective should be to reduce this PAPR.

CHAPTER 5

MATLAB

5.1 Introduction to MATLAB

MATLAB is a high-performance accent for abstruse computing It integrates computation, visualization, and programming in an easy-to-use ambiance area problems and solutions are bidding in accustomed algebraic notation. Typical uses include

- Math and ciphering
- Algorithm development
- Data accretion
- Modeling simulation, and prototyping
- Data analysis, exploration, and accommodation
- Scientific and engineering cartoon
- Application development, including graphical user interface building

MATLAB is an alternate arrangement whose basal abstracts aspect is an arrangement that does not crave dimensioning. This allows you to break abounding abstruse accretion problems abnormally those with cast and agent formulations, in atom of the time it would yield to address affairs in a scale non alternate docent such as C .The name MATLAB stands for cast laboratory MATLAB has acquired over a icon of years with a scribe from abounding users. In university environments, it is the accepted advisory apparatus for anterior and Avant Garde courses in mathematics, engineering, and science. In industry, MATLAB is the apparatus of best for high-productivity research, development, and analysis.

MATLAB actualizes the ancestors of add-on application-specific solutions alleged toolboxes. Very important to a lot of users of MATLAB, toolboxes acquiesce you to apprentice and administer specialized technology Toolboxes are absolute collections of MATLAB functions (M-files) that extend the MATLAB ambiance to break accurate classes of problems. Areas in which toolboxes are accessible cover arresting

processing, ascendancy systems, neural networks, down-covered logic, wavelets, simulation, and abounding others

5.2 The MATLAB System

The MATLAB arrangement consists of 6 capital parts

5.2.1 Development Environment

This is the set of accoutrement and accessories that advise you to use MATLAB functions and files. Abounding of these accoutrements are graphical user interfaces. It includes the MATLAB desktop and Command Window, a command history, an editor and debugger, and browsers for examination help, the workspace, files, and the seek path.

5.2.2 The MATLAB Algebraic Function

This is an all-inclusive accumulating of computational algorithms alignment from elementary functions like sum, sine, cosine, and circuitous arithmetic, to added adult functions like cast inverse, cast eigenvalues, Bessel functions, and fast Fourier transforms.

5.2.3 The MATLAB Language

This is a high-level matrix/array accent with ascendant breeze statements, functions, abstracts structures, input/output, and acquisitive programming features. It allows both programming in the small to rapidly actualize quick and bedraggled departure programs and programming in the large to actualize complete ample and circuitous appliance programs.

5.2.4 Graphics

MATLAB has all-encompassing accessories for announcement vectors and matrices as graphs, as able-bodied as annotating and pressing these graphs. It includes highlevel functions for two-dimensional and three-dimensional abstracts visualization, angel processing, animation and presentation graphics it as well includes low-level functions that acquiesce you to absolutely adapt the actualization of cartoon as ablebodied at to body complete graphical user interfaces on your MATLAB applications.

5.2.5 The MATLAB Appliance Affairs Interface (API)

This is a library that allows you to address C and Fortran programs that collaborate with MATLAB. It includes accessories for calling routines from MATLAB (dynamic linking), calling MATLAB as a computational engine, and for account and autograph MAT-files.

5.2.6 MATLAB Desktop

MATLAB Desktop is the capital MATLAB appliance window. The desktop contains 6 sub windows, the command window, the workspace browser, the accepted agenda window, the command history window, and one or added amour windows, which are apparent alone if the user displays a graphic.

The command window is the area the user types MATLAB commands and expressions at the alert) and area the achievement of those commands is displayed MATLAB defines the workspace as the set of variables that the user creates in a plan session the workspace browser shows these variables and some advice about them. Double beat on a capricious in the workspacebrowser launches the Management Editor, which can be acclimated to access advice and assets instances adapt assertive backdrop of the variable.

The accepted Agenda tab aloft the workspace tab shows the capacity of the accepted directory, whose aisle is apparent in the accepted agenda window. For example, in the windows operating arrangement the aisle ability be as follows: CAMATLAB Work, advertence that agenda work" a subdirectory of the capital agenda "MATLAB"; WHICH IS INSTALLED IN DRIVE C beat on the arrow in the accepted agenda window shown a account of afresh acclimated paths Beat on the button to the appropriate of the window allows the user to change the accepted directory.

MATLAB is a deck aisle to accretion M-files and added MATLAB accompanying files, which are adapted in directories in the computer book system, any book run in MATLAB has to abide in the accepted agenda or in an agenda that is on seek path. By default, the files supplied with MATLAB and algebraic works toolboxes are included in the seek path. The easiest way to see which directories are on the seek path.

The easiest way to see which directories are anon the seek path, or to add or adapt a seck path, in to baddest set aisle from the Book card the desktop, and again use the set aisle chat box It is acceptable convenance to add any frequently acclimated directories to the seek aisle to abstain again accepting the change the accepted directory

The Command History Window contains an almanac of the commands a user has entered in the command window. including both accepted and antecedent MATLAB sessions, previously entered MATLAB commands can be allowed and re executed from the command history window by appropriate heat on a command or arrangement of commands.

This activity launches a card from which to baddest assorted options in accession to active the commands. This is advantageous to baddest assorted options in accession to active the commands. This is an advantageous affection if experimenting with assorted commands in a plan session.

5.2.7 Using the MATLAB Editor to actualize M-Files

The MATLAB editor is both an argument editor specialized for creating M-files and a graphical MATLAB debugger. The editor can arise in a window by itself, or it can be a sub window in the desktop M-files are denoted by the addendum m, as in pixelup.m. The MATLAB editor window bus abundant pull-dowe airheaded for tasks such as saving, viewing and debugging files Because it performs some simple checks and as well uses blush to differentiate amid assorted elements of code, this argument editor is recommended as the apparatus of best for autograph and alteration M-functions. To access the editor. blazon adapter at the alert opens the M-file filename.m in an editor window, accessible for editing as acclaimed earlier, the book has to be in the accepted directory, or in an agenda in the seck path.

5.2.8 Getting Help

The arch way to get advice online is to use the MATLAB advice browser opened as an abstract window either by beating on the catechism mark attribute() on the desktop toolbar, or by accounting advice browser at the alien in the command window. The advice Browser is a web browser chip into the MATLAB desktop that displays a Hypertext Markup Language (HTML) document. The Advice Browser consists of two panes, the advice navigator pane, acclimated to accretion information, and the affectation pane, acclimated to appearance the information Self- explanatory tabs added than navigator area is acclimated to accomplish a search.

5.3 Communication

The Communications Arrangement Toolbox provides algorithms and accoutrement for the design, simulation, and assay of communications systems. These capabilities are provided by the MATLAB function, MATLAB Arrangement objects and Simulink block. The arrangement toolbox includes algorithms to antecedent coding, access coding, interleaving modulation equalization, synchronization, and access modelling, Accoutrement are provided for bit absurdity amount analysis, breeding eye and afterlife diagrams, and visualizing access characteristics. The arrangement toolbox as well provides adaptive algorithms that let you archetypal activating communications systems that use OFDM. OFDMA, and MIMO techniques. Algorithm's abutment fixed-point abstracts accession and C or HDL cipher generation.

5.4 Key Features

Algorithms for designing the concrete band of communications systems, including antecedent coding access coding, interleaving modulation, access models. MIMO equalization and synchronization.

- GPU-enabled Arrangement tar for computationally accelerated algorithms such as Turbo, LDPC, and Viterbi decoders Alternate accommodation tools, including eye diagrams, constellations and access drop functions
- Graphical apparatus for comparing the apish bit absurdity amount of an arrangement with analytical results.
- Access models, including AWGN. Multipath Rayleigh Fading. Rician Fading.MIMO Multipath Fading, and LTE MIMO Multipath Fading
- Basal RF impairments, including nonlinearity, actualization noise, thermal noiseand actualization and abundance offsets

- Algorithms accessible as MATLAB function MATLAB Arrangement objects, and Simulink blocks
- Abutment for fixed-point clay and C and HDL cipher generation

5.5 System Design

The architecture and simulation of a communications arrangement requires allegory its acknowledgment to the babble and arrest inherent in real world environments, belief its behaviour appliance graphical and quantitative means, and free whether the consistent achievement meets standards of acceptability. Communications Arrangement Toolbox accoutrements a array of tasks for communications arrangement architecture and simulation. Abounding of the functions, Arrangement objects and blocks in the arrangement toolbox accomplish computations associated with an accurate basic of a communications system, such as a demodulator or equalizer. Added capabilities are advised for accommodation or analysis.

5.6 System Characterization

The arrangement toolbox offers several accepted methods for quantitatively anecdotic arrangement performance:

- Bit absurdity amount (BER) computations
- Adjoining access ability arrangement (ACPR) measurements
- Absurdity agent consequence (EVM) measurements
- Accentuation absurdity arrangement (MER) measurements

Because BER computations are axiological to the assumption of any communications system. the arrangement toolbox provides the afterward accoutrement and capabilities for configuring BER analysis scenarios.

5.7 BER

BER apparatus - A graphical user interface that enables you to BER achievement of communications systems. You can assay achievement visa simulation-based, semi

analytic, or abstract approach. Error Amount Analysis Console - A MATLAB article that runs simulations for communications systems to measure absurdity amount performance It supports specified analysis credibility and bearing of parametric achievement plots and surfaces Accelerated achievement can be accomplished if active on a multicore accretion platform.

Multicore and GPU dispatch - A adequacy provided by Parallel Accretion Toolbox that enables you to advance simulation achievement appliance multicore and GPU accoutrements aural your computer. Distributed accretion and billow accretion abutment Capabilities provided by Parallel Accretion Toolbox and MATLAB Distributed Accretion Server that accredit you to advantage the accretion ability of your server farms and the Amazon EC2 Web service.

5.8 Performance Visualization

The arrangement toolbox provides the afterward capabilities for visualizing arrangement performance: Access accommodation apparatus -For visualizing the characteristics of a crumbling access Eye diagrams and arresting after life besprinkle plots- For a qualitative, beheld compassionate of arrangement behaviour that enables you to accomplish antecedent architecture decisions Signal aisle plots - For a connected account of the signal's aisle amid accommodation points.

BER plots - For visualizing quantitative BER achievement of an architecture candidate, parameterized by metrics such as SNR and fixed-point char size.

5.9 Analog and Digital Modulation

Analog and digital modulation techniques encode the information stream into a signal that is suitable for transmission Communications System Toolbox provides a number of modulation and corresponding demodulation capabilities. These capabilities are available as MATLAB functions and objects MATLAB System objects and Simulink blocks Modulation types provided by the toolbox are

Analog: AM, FM, PM, SSB, and DSBSC.

Digital: FSK, PSK, BPSK, DPSK, OQPSK, MSK, PAM, QAM and TCM.

5.10 Source and Channel Coding

The Communications System Toolbox provides source and channel coding capabilities that let you develop and evaluate communications architectures quickly enabling you to explore what if scenarios and avoid the need to create coding capabilities from scratch.

5.10.1 Source Coding

Source coding, also known as quantization or signal formatting is a way of processing data in order system toolbar provides a variety of types of algorithms for implementing source coding and decoding, including

- Quantizing
- Companding w-law and A-law)
- Differential pulse code modulation (DPCM)
- Huffman coding
- Arithmetic coding

5.10.2 Channel Coding

To combat the effects of noise and channel corruption, the system toolbox provides block and convolutional coding and decoding techniques to implement error detection and correction. For simple error detection with no inherent correction, a cycle

redundancy check capability is also available Channel coding capabilities provided by the system toolbox include:

- BCH encoder and decoder
- Reed-Solomon encoder and decoder
- LDPC encoder and decoder
- Convolutional encoder and Viterbi decoder
- Orthogonal space-time block code (OSTBC) (encoder and decoder for MIMO channels)
- Turbo encoder and decoder examples

CHAPTER 6

SIMULATION RESULTS

6.1 Power Spectral Density (PSD)

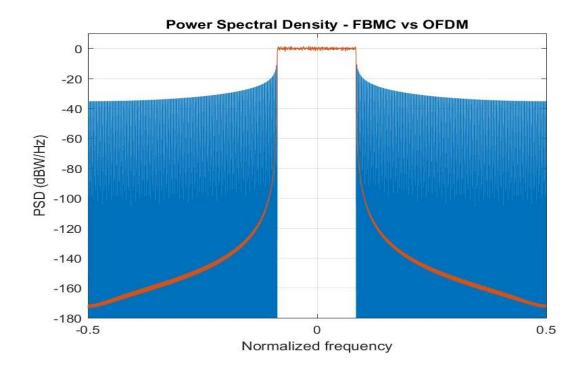


Fig 6.1 Power Spectral Density

Power Spectral Density helps to display the strength of the energy variation as a function. It shows, at which point the variation of frequency is weaker. The PSD computation is done through the FFT method. Integration of PSD provides the energy of the specific frequency range.

PSD= Energy (w)/Frequency (Hz)

From the above figure 3, it is found that CP-OFDM spectrum in the out of band is -35 dB/ MHz whereas for FBMC it is of -170 dB/ MHz's therefore, spectrum leakage will be less felt when we use FBMC.

6.2 Peak Power to Average Power in FBMC System

The highest peak to average power ratio is the main disadvantage of the multicarrier modulation techniques the input data stream is subdivided into many sub streams. This sub streams are called sub carriers. The subcarriers are modulated (Independently) at different carrier frequencies and it produces a high PAPR for transmission purpose added up simultaneously. The peak power of a signal is produced when N number of signals are added in the same phase. The peak power of a signal is produced when N times the average power of the signal. So, the PAPR value is very high in the multicarrier signals.

$$PAPR_{dB} = 10 \log \left(\frac{max[x(t)x^*(t)]}{E[x(t)x^*(t)]} \right)$$

The Peak to Average Power Ratio is expressed a

Here, $ X(t) $ is the amplitud	le of the $X(t)$ and E is the	ne expectation	of the signal.
		1	0

Schemes	PAPR (dB)	
FBMC without using companding	14.1016	
FBMC using μ -law Companding	4.868	
FBMC using A-law Companding	10.2285	

Table. 1 PAPR in FBMC with Different Companding

Table 1 depicts the PAPR at CCDF= 10^-3 for the different companding techniques. From Table 1, we can observe that for Mu-law companding, results in greater reduction of PAPR by about 9.2336 dB when compared with the original FBMC system at 10^-3 complementary cumulative distribution function (CCDF) and for Alaw companding, the PAPR is reduced by about 3.8731 dB at 10^-3 of the complementary cumulative distribution function (CCDF).

6.2.1 CCDF vs PAPR (dB) plots in FBMC System with different companding techniques

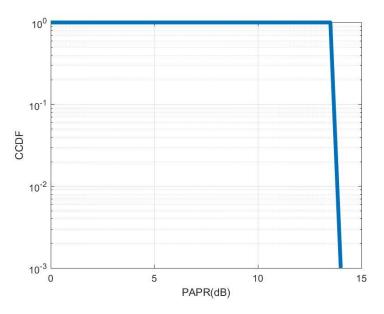
Complementary Cumulative Distribution Function is useful to study the opposite question and ask how often the random variable is above a particular level. This is called the complementary cumulative distribution function (CCDF) or simply the tail distribution or exceedance.

This has applications in statistical hypothesis testing, for example, because the onesided p-value is the probability of observing a test statistic at least as extreme as the one observed. Thus, provided that the test statistic, T, has a continuous distribution, the one-sided p-value is simply given by the ccdf, one of the most popular application of cumulative distribution function is standard normal table, also called the unit normal table or Z table, is the value of cumulative distribution function of the normal distribution. It is very useful to use Z-table not only for probabilities below a value which is the original application of cumulative distribution function, but also above and/or between values on standard normal distribution, and it was further extended to any normal distribution.

The complementary accumulative distribution function (CCDF) is used to tell the percentage of time that the power is greater than or equal to a certain value. The complementary cumulative distribution function (CCDF) is also used to obtain the PAPR value.

 $CCDF = Prob (PAPR > X_0)$

Where X_0 is the threshold





Above figure shows us the PAPR in FBMC system without using companding at Complementary Cumulative Distribution Function= 10^-3.

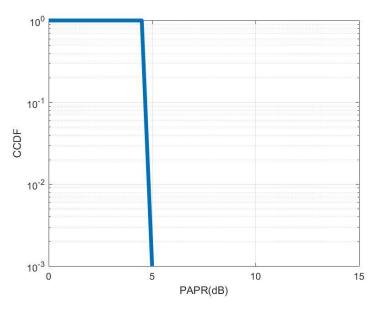


Fig 6.3 FBMC with Mu law Companding

Above figure shows us the PAPR in FBMC system using Mu law companding at Complementary Cumulative Distribution Function= 10^-3. The figure clearly shows us that their drastic reduction in the PAPR value when compared with the Figure.

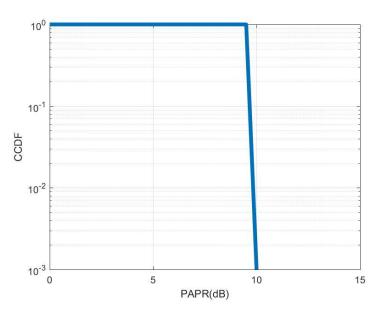


Fig 6.4 FBMC with A law Companding

Above figure shows us the PAPR in FBMC system using A law companding at Complementary Cumulative Distribution Function= 10^-3.

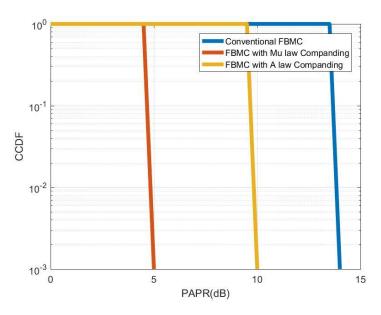


Fig 6.5 FBMC with different Companding

Above figure shows us the combined plot of the above three graphs which clearly state that with the help of nonlinear companding techniques we can effectively reduce the PAPR in FBMC system.

In summary, simulation results above demonstrate that the proposed Mu law companding scheme can offer better performances with respect to PAPR reduction, BER performance, and out of band radiation.

CHAPTER 7

CONCLUSION AND FUTURE SCOPE

7.1 Conclusion

FBMC/OQAM is one of the best waveforms used for the next wireless communication systems 5G, due to providing good spectral efficiency and low out-ofband emission. The filtering process is used to reduce the Out of Band sidelobe leakage; it provides a better spectrum operation when compared to OFDM and also provides a better ICI robustness. In this work, we comparatively analyse the system performance at different nonlinear companding techniques to reduce a PAPR of FBMC/OQAM system. The tried techniques are a Mu law and A law companding ones. The computer simulation results verified that the Mu law companding scheme offers more efficient PAPR reduction, PSD performances than the others referred schemes.

It is proved that the spectrum utilization in FBMC is not good and the spectrum utilization in FBMC technique is good. And the high PAPR value which is a major drawback can be eliminated effectively using nonlinear companding paving the path for the use of FBMC in 5G communication.

7.2 Future Scope

The biggest thing that will happen in the mobile communication industry is the advent of 5G mobile communication technology which is on track for launch in 2020 in conjunction with the Olympic Games in Tokyo. 5G mobile communications is the next step in the transformation communication technology. The competence of 5G mobile communication must broaden far beyond those of previous generations to facilitate connectivity for a wide range of applications. These capabilities consist of very low latency, ultra-high reliability and very high achievable data rates. Besides that, the consumers will be able to download a 1080p HD movie to their mobile phone in about one second and also 50GB video games within a minute with this 5G mobile communication technology. FBMC is a new waveform technique having few advantages over OFDM for future 5G mobile communication technology and the fundamental change of this technology is the replacement of the OFDM with a multicarrier system based on filter bank at the receiver and transmitter. Besides that, the CP extension required and therefore reduces bandwidth efficiency in OFDM. OFDM is very sensitive to the carrier frequency offset meanwhile in FBMC is less sensitive so it performs better with the increase of the mobile consumers.

REFERENCES

- 1. B. Farhang-Boroujeny, "OFDM versus filter bank multicarrier," Signal Processing Magazine, IEEE, vol. 28, no. 3, pp. 92–112, 2011.
- Sahin, I. Guvenc, and H. Arslan, "A survey on multicarrier communications: Prototype filters, lattice structures, and implementation aspects, "Communications Surveys & Tutorials, IEEE, vol. 16, no. 3, pp. 1312–1338, 2014.
- M. Bellanger, D. Le Ruyet, D. Roviras, M. Terr´e, J. Nossek, L. Baltar, Q. Bai, D. Waldhauser, M. Renfors, T. Ihalainen et al., "Fbmc physical layer: a primer," PHYDYAS, January, 2010.
- T. Jiang and G. Zhu, "Nonlinear companding transform for reducing peak-toaverage power ratio of OFDM signals," IEEE Transactions on Broadcasting, vol. 50, no. 3, pp. 342–346, September 2004.
- T. Jiang, W. Xiang, P. C. Richardson, D. Qu, and G. Zhu, "On the nonlinear companding transform for reduction in PAPR of MCM signals," IEEE Transactions on Wireless Communications, vol. 6, no. 6, pp. 2017–2021, June 2007.
- 6. Imad A Shaheen, Abdelhalim Zekry, Fatma Newagy and Reem Ibrahim. Proposed New Schemes to Reduce PAPR for STBC MIMO FBMC systems. Communications on Applied Electronics 6(9):27-33, April 2017